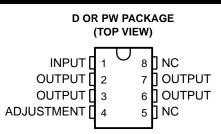
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- Output Voltage Range Adjustable From 1.2 V to 32 V When Used With an External Resistor Divider
- Output Current Capability of 100 mA
- Input Regulation Typically 0.01% Per Input-Voltage Change
- Output Regulation Typically 0.5%
- Ripple Rejection Typically 80 dB

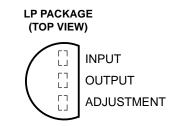
### description

The TL317 is an adjustable three-terminal positive-voltage regulator capable of supplying 100 mA over an output-voltage range of 1.2 V to 32 V. It is exceptionally easy to use and requires only two external resistors to set the output voltage.

In addition to higher performance than fixed regulators, this regulator offers full overload



NC – No internal connection OUTPUT terminals are all internally connected.



protection available only in integrated circuits. Included on the chip are current-limiting and thermal-overload protection. All overload-protection circuitry remains fully functional, even when ADJUSTMENT is disconnected. Normally, no capacitors are needed unless the device is situated far from the input filter capacitors, in which case an input bypass is needed. An optional output capacitor can be added to improve transient response. ADJUSTMENT can be bypassed to achieve very high ripple rejection, which is difficult to achieve with standard three-terminal regulators.

In addition to replacing fixed regulators, the TL317 regulator is useful in a wide variety of other applications. Since the regulator is floating and sees only the input-to-output differential voltage, supplies of several hundred volts can be regulated as long as the maximum input-to-output differential is not exceeded. Its primary application is that of a programmable output regulator, but by connecting a fixed resistor between ADJUSTMENT and OUTPUT, this device can be used as a precision current regulator. Supplies with electronic shutdown can be achieved by clamping ADJUSTMENT to ground, programming the output to 1.2 V, where most loads draw little current.

The TL317C is characterized for operation over the virtual junction temperature range of 0°C to 125°C.

AVAILABLE OPTIONS					
	PACKAGED DEVICES				
ТJ	SMALL OUTLINE (D)	PLASTIC (LP)	PLASTIC THIN SHRINK SMALL OUTLINE (PW)		
0°C to 125°C	TL317CD	TL317CLP	TL317CPW		

The D and LP packages are available taped and reeled. Add the suffix R to device type (e.g., TL317CDR). The PW package is only available taped and reeled.



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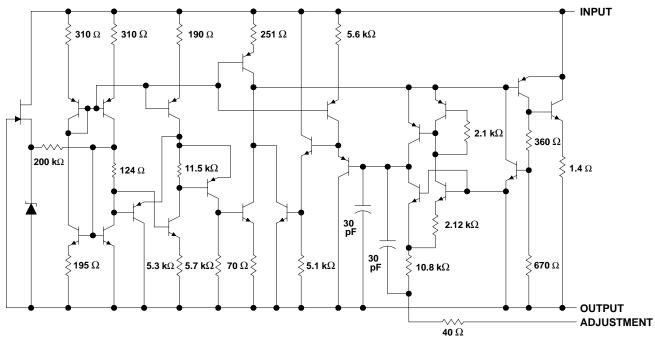
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## schematic



NOTE A: All component values shown are nominal.

# absolute maximum ratings over operating temperature range (unless otherwise noted)<sup>†</sup>

Input-to-output differential voltage, $V_{I} - V_{O}$		35 V
Package thermal impedance, $\theta_{JA}$ (see Notes 1 and 2):	D package	97°C/W
	LP package	156°C/W
	PW package	149°C/W
Lead temperature 1,6 mm (1/16 inch) from case for 10	seconds	260°C
Storage temperature range, T <sub>sto</sub>		–65°C to 150°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. Maximum power dissipation is a function of  $T_J(max)$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any allowable ambient temperature is  $P_D = (T_J(max) - T_A)/\theta_{JA}$ . Operating at the absolute maximum  $T_J$  of 150°C can affect reliability.

2. The package thermal impedance is calculated in accordance with JESD 51-7.

## recommended operating conditions

			MIN	MAX	UNIT
$V_I - V_O$	Input-to-output voltage differential			35	V
IO	Output current		2.5	100	mA
Тј	Operating virtual-junction temperature	TL317C	0	125	°C



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# electrical characteristics over recommended operating virtual-junction temperature range (unless otherwise noted)

DADAMETED	TEST CONDITIONS <sup>†</sup>		TL317C			
PARAMETER			MIN	TYP	MAX	UNIT
Input voltage regulation (see Note 3)	$V_{I} - V_{O} = 5 V$ to 35 V	T <sub>J</sub> = 25°C		0.01	0.02	<b>- %</b> ∨
		I <sub>O</sub> = 2.5 mA to 100 mA		0.02	0.05	
	V <sub>O</sub> = 10 V,	f = 120 Hz		65		
Ripple regulation	$V_{O}$ = 10 V, 10-µF capacitor between ADJUSTMENT and ground		66	80		dB
	$V_{I} = 5 V \text{ to } 35 V,$ $I_{O} = 2.5 \text{ mA to } 100 \text{ mA},$ $T_{J} = 25^{\circ}\text{C}$	$V_{O} \le 5 V$		25		mV
Output voltage regulation		$V_{O} \ge 5 V$		5		mV/V
	$V_{I} = 5 V \text{ to } 35 V,$ IO = 2.5 mA to 100 mA	$V_{O} \le 5 V$		50		mV
		$V_{O} \ge 5 V$		10		mV/V
Output voltage change with temperature	T <sub>J</sub> = 0°C to 125°C			10		mV/V
Output voltage long-term drift	After 1000 hours at T <sub>J</sub> = 125°C and V <sub>I</sub> – V <sub>O</sub> = 35 V			3	10	mV/V
Output noise voltage	f = 10 Hz to 10 kHz,	TJ = 25°C		30		μV/V
Minimum output current to maintain regulation	V <sub>I</sub> – V <sub>O</sub> = 35 V			1.5	2.5	mA
Peak output current	$V_I - V_O \le 35 V$		100	200		mA
ADJUSTMENT current				50	100	μA
Change in ADJUSTMENT current	$V_{I} - V_{O} = 2.5 V \text{ to } 35 V,$	I <sub>O</sub> = 2.5 mA to 100 mA		0.2	5	μA
Reference voltage (output to ADJUSTMENT)	$V_I - V_O = 5 V \text{ to } 35 V,$ P $\leq$ rated dissipation	$I_{O} = 2.5 \text{ mA to } 100 \text{ mA},$	1.2	1.25	1.3	V

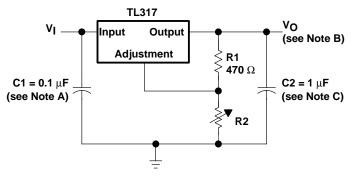
<sup>†</sup> Unless otherwise noted, these specifications apply for the following test conditions:  $V_I - V_O = 5$  V and  $I_O = 40$  mA. Pulse-testing techniques must be used that maintain the junction temperature as close to the ambient temperature as possible. All characteristics are measured with a 0.1- $\mu$ F capacitor across the input and a 1- $\mu$ F capacitor across the output.

NOTE 3: Input voltage regulation is expressed here as the percentage change in output voltage per 1-V change at the input.



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# **APPLICATION INFORMATION**



- NOTES: A. Use of an input bypass capacitor is recommended if regulator is far from the filter capacitors.
  - B. Output voltage is calculated from the equation:  $y_{1} = y_{2} (1 + R^{2})$

$$V_{O} = V_{ref} \left( 1 + \frac{RT}{R1} \right)$$

where: V<sub>ref</sub> equals the difference between OUTPUT and ADJUSTMENT voltages ( $\approx$ 1.25 V).

C. Use of an output capacitor improves transient response but is optional.

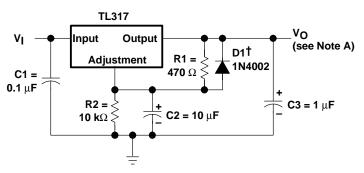
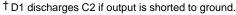
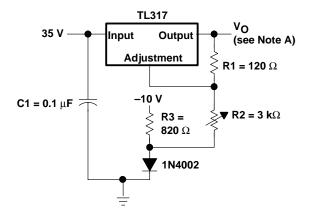


Figure 1. Adjustable Voltage Regulator



NOTE A: Use of an output capacitor improves transient response but is optional.

## Figure 3. Regulator Circuit With Improved Ripple Rejection



NOTE A: Output voltage is calculated from the equation:

$$V_{\rm O} = V_{\rm ref} \left( 1 + \frac{R2 + R3}{R1} \right) - 10$$

where:  $V_{ref}$  equals the difference between OUTPUT and ADJUSTMENT voltages ( $\approx$ 1.25 V).

# Figure 2. 0-V to 30-V Regulator Circuit

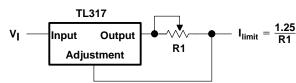


Figure 4. Precision Current-Limiter Circuit



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## **APPLICATION INFORMATION**

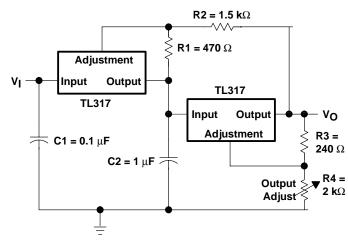
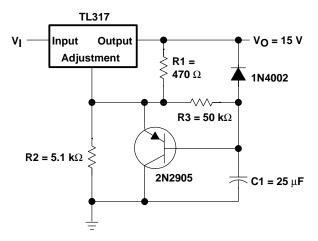


Figure 5. Tracking Preregulator Circuit





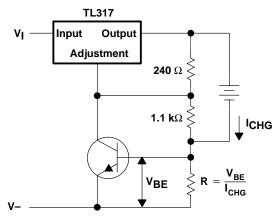


Figure 8. Current-Limited 6-V Charger



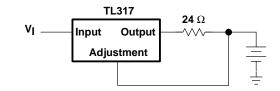
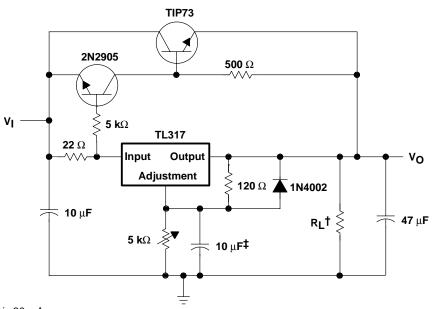


Figure 7. 50-mA Constant-Current Battery-Charger Circuit

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# **APPLICATION INFORMATION**

<sup>†</sup> Minimum load current is 30 mA.

<sup>‡</sup> Optional capacitor improves ripple rejection.

# Figure 9. High-Current Adjustable Regulator



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